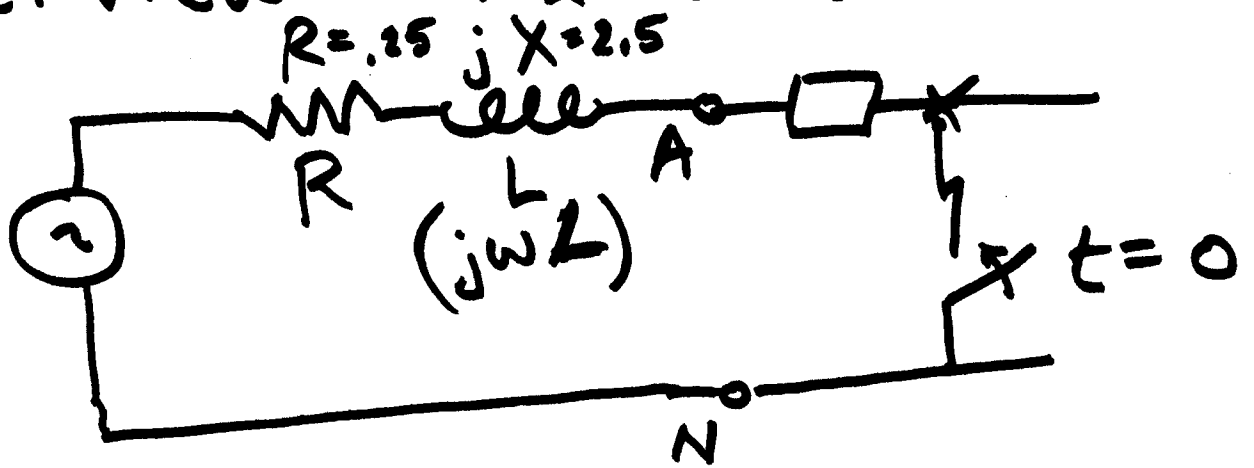


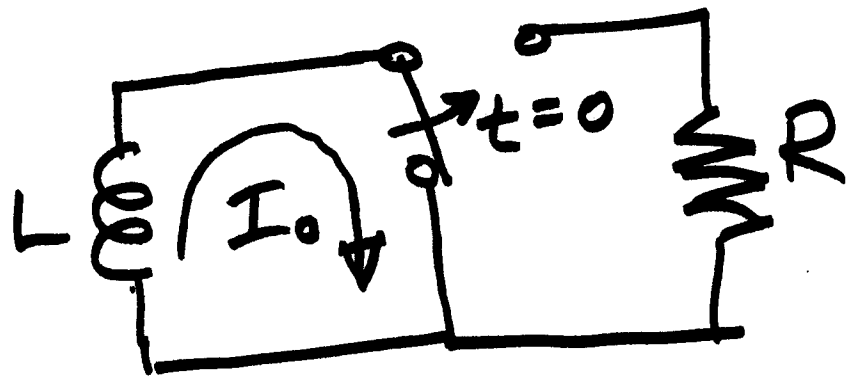
Topics for Today:

- Startup
 - Web page: <http://www.ee.mtu.edu/faculty/bamork/ee5220/>
 - Book, references, syllabus, more are on web page.
 - Software - ATP/EMTP, Matlab
 - EE5220-L@mtu.edu (participation = min half letter grade)
 - Lectures - new videostreams, some archived videos also
 - Daily lecture notes scanned and .pdf file archived
 - Exercises posted as pdf on Canvas.
 - Grading: grad students must achieve BC (75%) or higher.
 - Prereqs: - Circuit Analysis RLC Responses, EE5200
 - Do all exercises in Ch.1 (solutions are posted) note typo.
- **Chapters 1 and 2, probs 1.2, 1.3, 2.2, 2.3, 2.7 due next Wed.**
-
- Get your textbook! Dig out your old Circuits book!
- Review of Laplace, RLC circuits
- Basic use of ATP (prereq/tutorials from EE5200), see <http://www.ee.mtu.edu/faculty/bamork/ee5200/>

Overview - RL Circuits



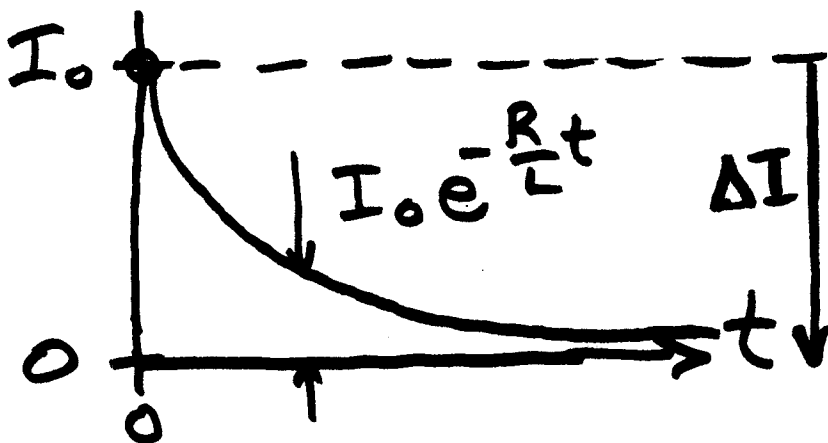
BASIC:



1) Identify Initial & Final State

$$i_L(0) = I_0 = i_L^- = i_L^+$$

$$i_L(\infty) = 0$$



$$\begin{aligned} i_L(t) &= I_\infty - \underline{\Delta I} e^{-\frac{R}{L}t} \\ &= I_\infty - (I_\infty - I_0) e^{-\frac{R}{L}t} \\ &= 0 - (0 - I_0) e^{-\frac{R}{L}t} \\ &= +I_0 e^{-\frac{R}{L}t} \end{aligned}$$

RL time constant

$$\tau = \frac{L}{R} \text{ s}$$

Consider $e^{-\frac{t}{\tau}} = e^{-\frac{R}{L}t}$

Increase $R \rightarrow$ Faster rate of change, less time to final state.

Increase $L \rightarrow$ Slower rate of change, longer to final state.

Stored energy:

$$= \frac{1}{2} Li^2 \text{ Joule}$$

$$= \frac{1}{2} Cv^2$$

$$= \frac{1}{2} kx^2 \text{ (spring)}$$

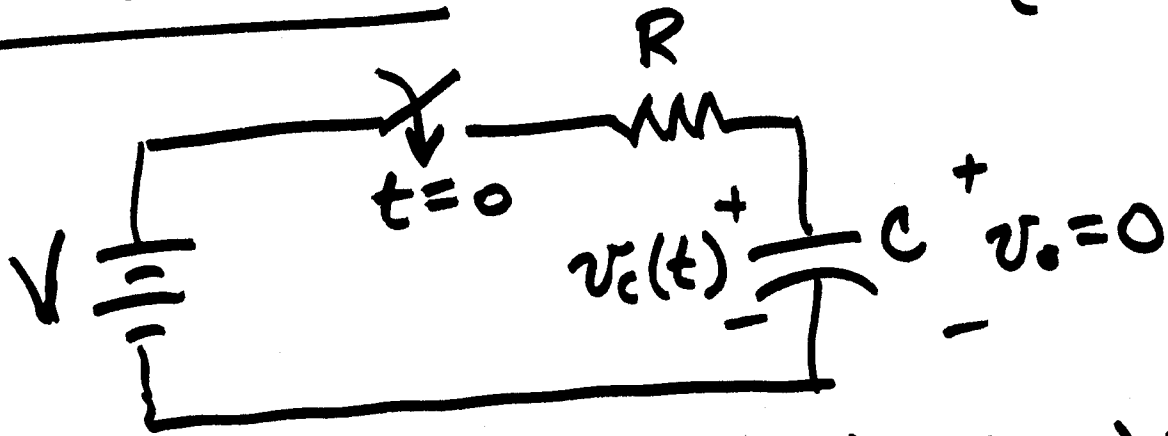
$$= \frac{1}{2} I\omega^2 \text{ (fly-wheel)}$$



"H"

RC Circuits -

$$\tau = RC \text{ s}$$



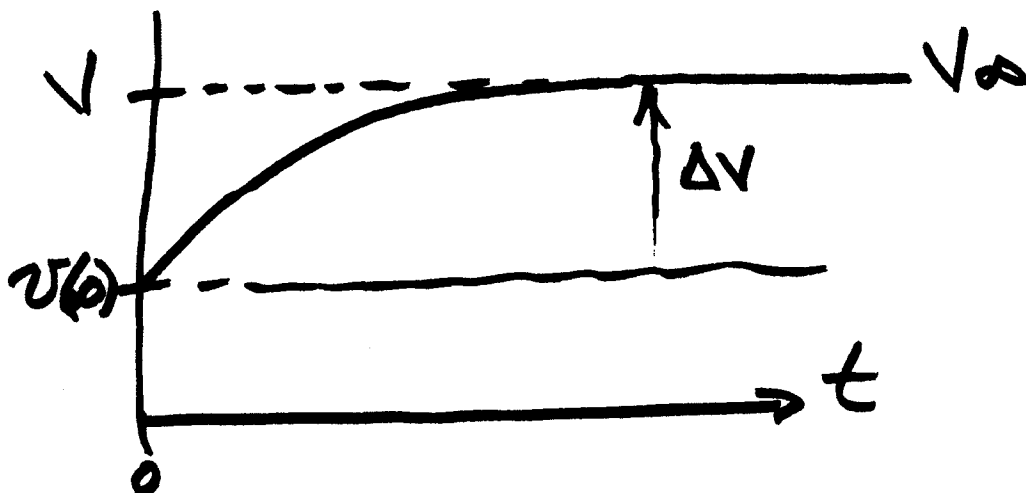
Initially: $v_c(0) = v_c(0^+) = v_c(\infty) = 0$

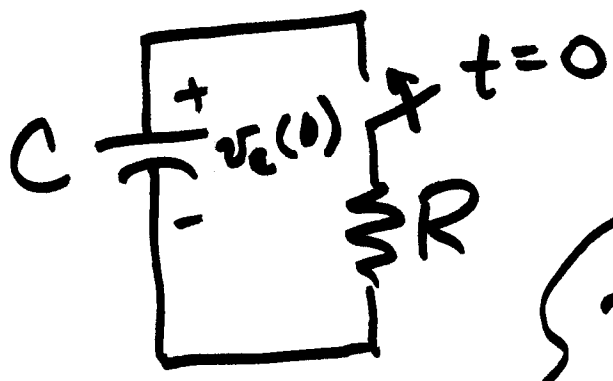
Final State: $v_c(\infty) = V$

$$\Delta V_c = V_\infty - V_0 = V - V_0$$

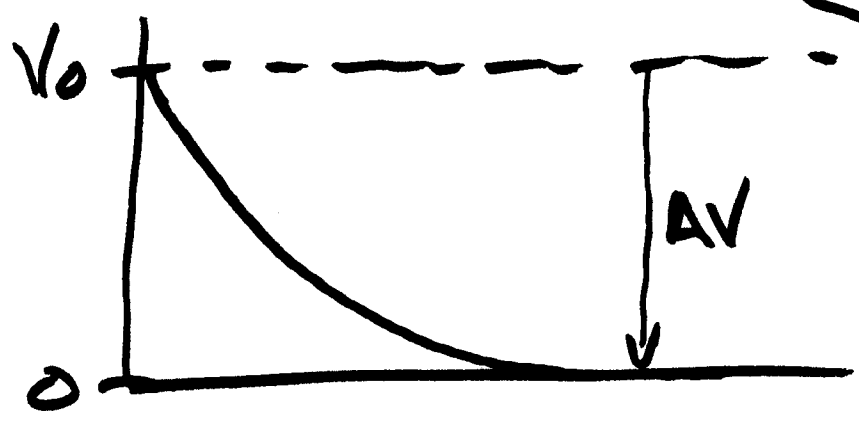
Note: ΔV_c drives response!

$$v_c(t) = V - \Delta V e^{-\frac{t}{RC}}$$
$$= \boxed{v_c(\infty) - \Delta V e^{-\frac{t}{RC}}}$$





$v_c(0) = V_0$
 $v_c(\infty) = 0$
 $\Delta V = \underbrace{V_0 - 0}_{V_0}$



$$\begin{aligned}
 v_c(t) &= v_c(\infty) - \frac{\Delta V e^{-\frac{t}{RC}}}{1} \\
 &= v_c(0) e^{-\frac{t}{RC}}
 \end{aligned}$$

time constants

Compare τ to 60-Hz period

$\frac{1.5}{.25} = \frac{X}{R} \approx 10$ for Transmission System

$\tau = \frac{L}{R} = \frac{X}{\omega R} = \frac{2.5}{377 \times .25} \approx \del{0.028} 0.028 \text{ s}$

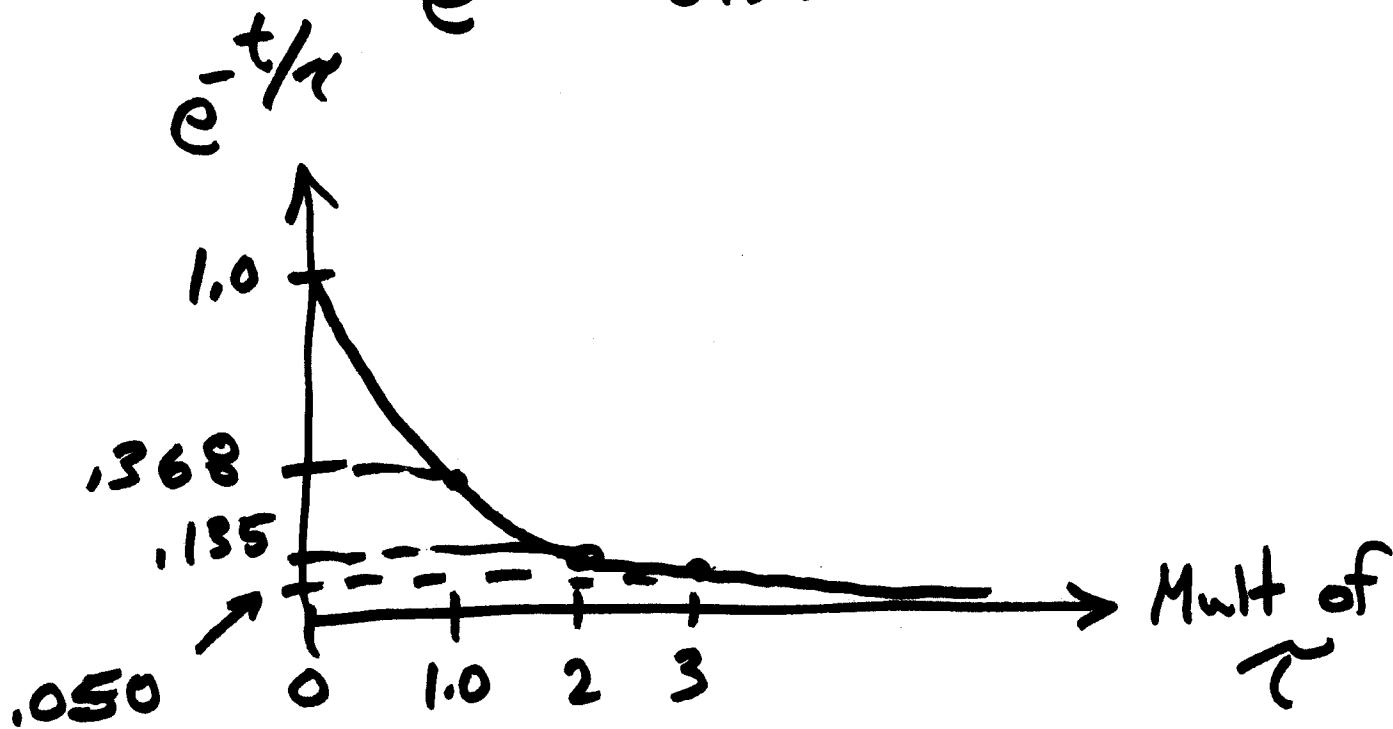
~~0~~ms \Leftrightarrow $\frac{1}{60} \text{ s} = \underline{16.67 \text{ ms}}$

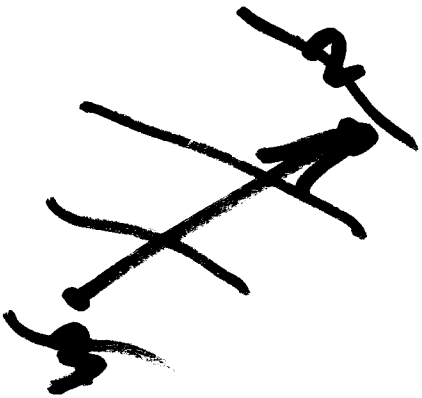
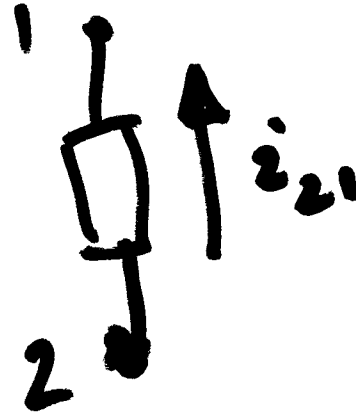
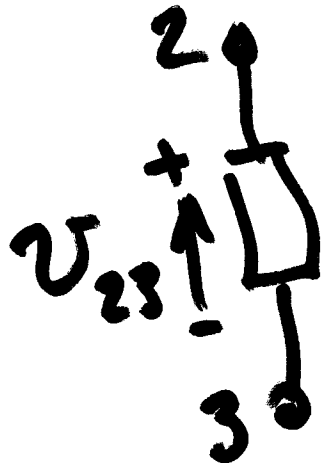
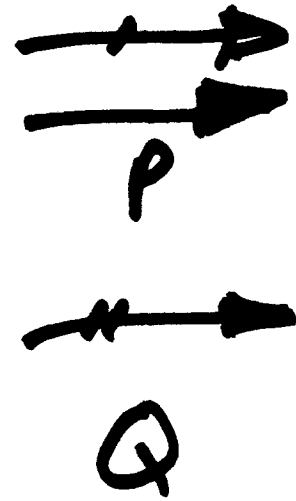
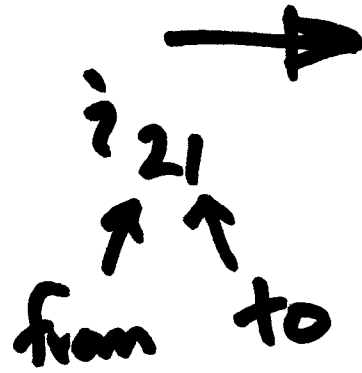
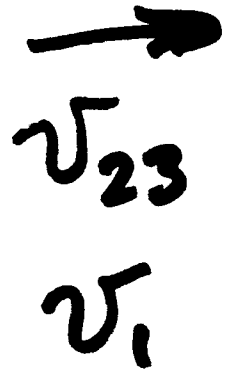
28ms

$$e^{-\frac{t}{\tau}}$$

At one time constant

$$e^{-1} = 0.368$$





The following general solutions may be used to calculate the transient response of an RLC circuit. The values of α and ω_0 must first be found and used to determine whether response is overdamped, underdamped, or critically damped.

RLC NATURAL RESPONSE - Parallel or series: $v_C(t)$, $i_C(t)$, $v_L(t)$, $i_L(t)$

$f(t) = A_1 e^{s_1 t} + A_2 e^{s_2 t}$	Overdamped
$f(t) = e^{-\alpha t} (B_1 \cos \omega_d t + B_2 \sin \omega_d t)$	Underdamped
$f(t) = D_1 t e^{-\alpha t} + D_2 e^{-\alpha t}$	Critically Damped

RLC STEP RESPONSE - Parallel or series: $v_C(t)$, $i_C(t)$, $v_L(t)$, $i_L(t)$

$f(t) = f(\infty) + A_1 e^{s_1 t} + A_2 e^{s_2 t}$	Overdamped
$f(t) = f(\infty) + e^{-\alpha t} (B_1 \cos \omega_d t + B_2 \sin \omega_d t)$	Underdamped
$f(t) = f(\infty) + D_1 t e^{-\alpha t} + D_2 e^{-\alpha t}$	Critically Damped

Coefficients for a particular case are determined by evaluating initial conditions. The expressions for each general solution and its derivative evaluated at $t=0^+$ are given below.

OVERDAMPED:

$$f(0^+) = A_1 + A_2 \quad [+ f(\infty) \text{ if step response}]$$

$$\frac{d f(0^+)}{dt} = s_1 A_1 + s_2 A_2$$

UNDERDAMPED:

$$f(0^+) = B_1 \quad [+ f(\infty) \text{ if step response}]$$

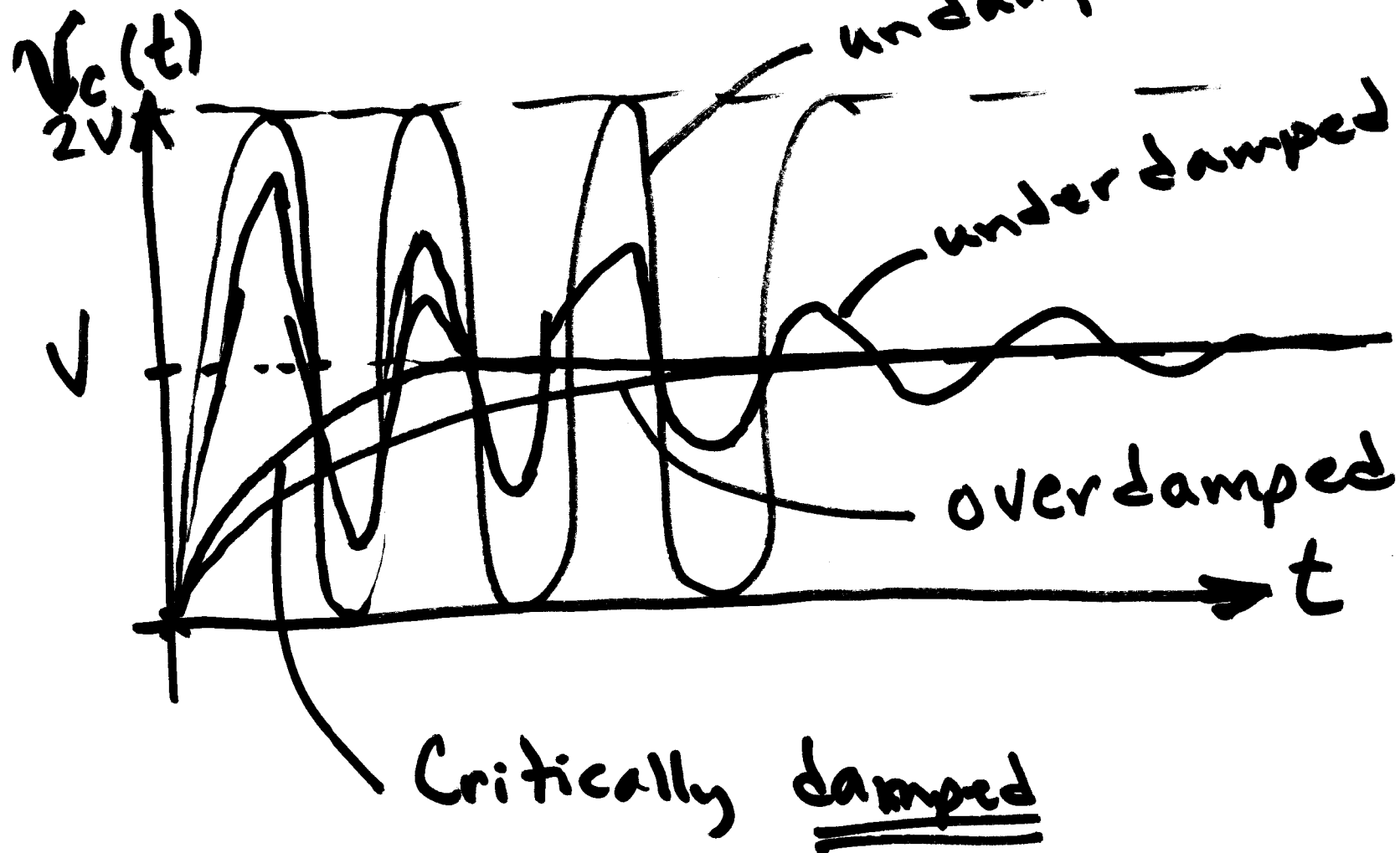
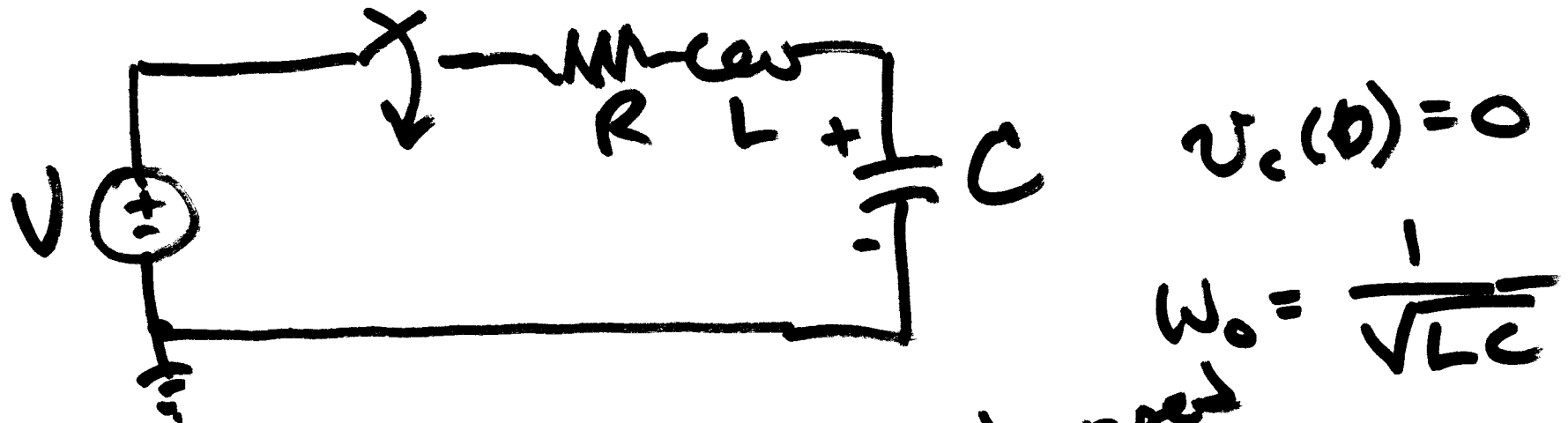
$$\frac{d f(0^+)}{dt} = -\alpha B_1 + \omega_d B_2$$

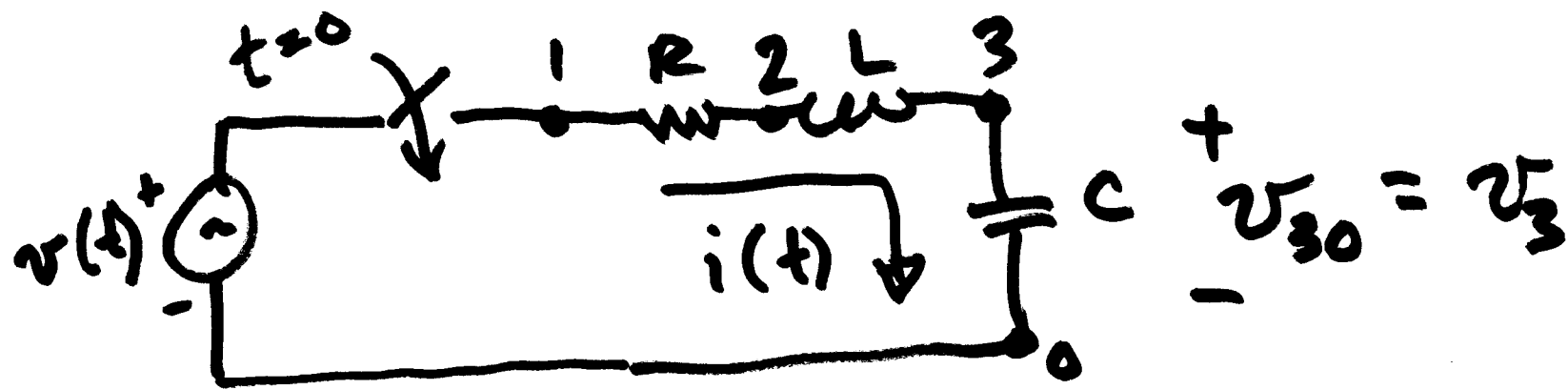
CRITICALLY DAMPED:

$$f(0^+) = D_2 \quad [+ f(\infty) \text{ if step response}]$$

$$\frac{d f(0^+)}{dt} = D_1 - \alpha D_2$$

The numeric initial values of the function and its derivative at $t=0^+$ are obtained from your knowledge of the particular circuit: parameter values, initial conditions, type of switching occurring, and differential relationships of current and voltage in the inductor and capacitor at $t=0^+$.

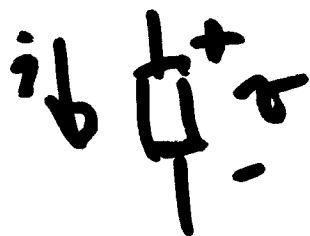




- Diff. Eqs - Handout
- Laplace
- Intuitive methods.

Review of CKts! - EE5200

- Subscript notations
- Active vs. passive sign conv.
- P, Q flow
- Phase angles



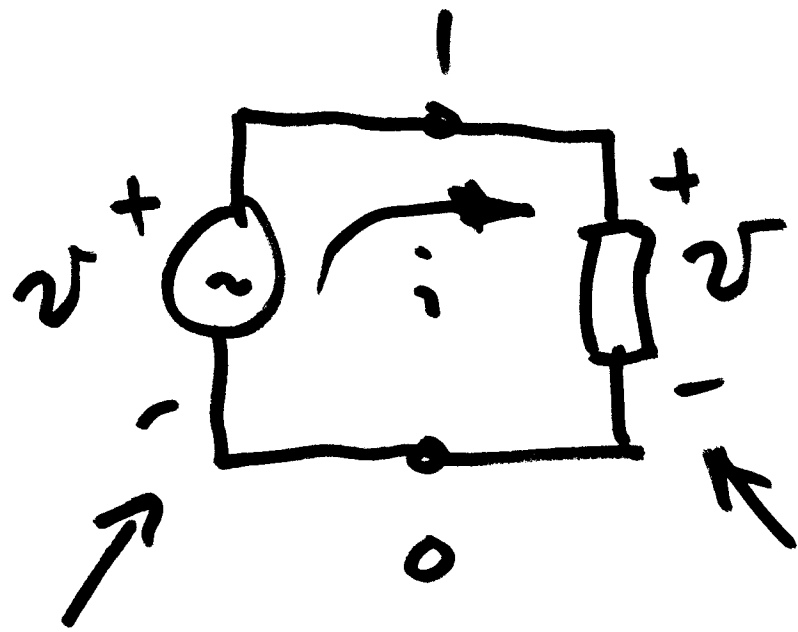
$$p(t) = v(t)i(t)$$

$$= P \angle \theta$$

v_{23}
 v_1

i_{21}
from to

P
 Q



$$P = vi = P_{out}$$

$$P = vi = P_{in}$$

APPENDIX A

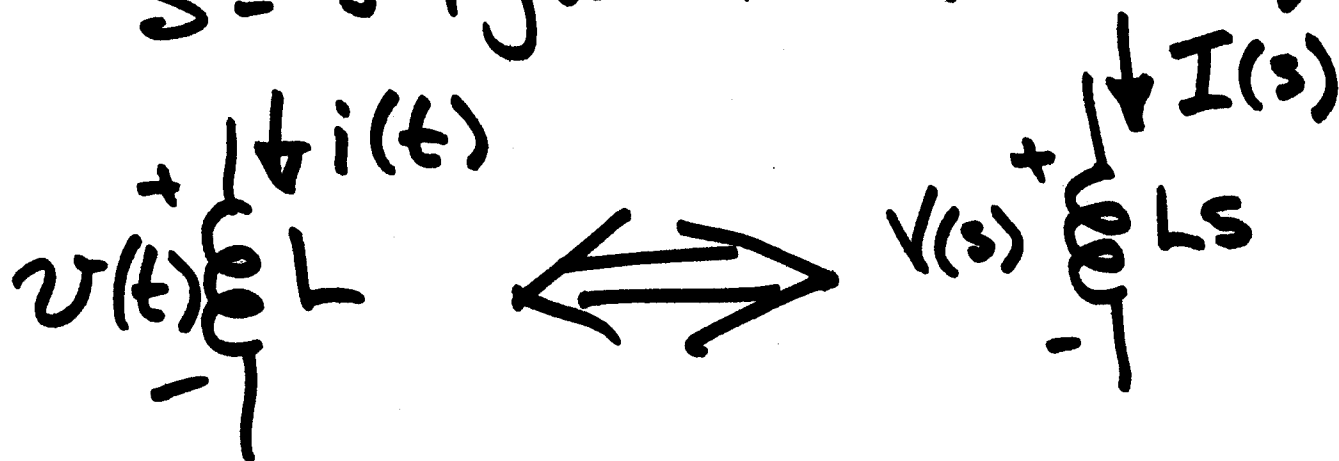
LAPLACE TRANSFORM TABLE

Laplace transform, $F(s)$	Time function, $f(t)$, $t \geq 0$
1	$\delta(t_0)$, unit impulse at $t = t_0$
$\frac{1}{s}$	$U(t)$, unit step function
$\frac{1}{s^2}$	t
$\frac{2}{s^3}$	t^2
$\frac{n!}{s^{n+1}}$	t^n
$\frac{1}{s+a}$	e^{-at}
$\frac{1}{(s+a)(s+b)}$	$\frac{e^{-at} - e^{-bt}}{b-a}$
$\frac{1}{(s+a)^n}$	$\frac{1}{(n-1)!} t^{n-1} e^{-at}$
$\frac{s+\alpha}{(s+a)(s+b)}$	$\frac{1}{(b-a)} [(\alpha-a)e^{-at} - (\alpha-b)e^{-bt}]$
$\frac{1}{(s+a)(s+b)(s+c)}$	$\frac{e^{-at}}{(b-a)(c-a)} + \frac{e^{-bt}}{(c-b)(a-b)} + \frac{e^{-ct}}{(a-c)(b-c)}$
$\frac{\omega}{s^2 + \omega^2}$	$\sin \omega t$
$\frac{s}{s^2 + \omega^2}$	$\cos \omega t$
$\frac{\omega}{(s+\omega)^2}$	$\omega t e^{-\omega t}$
$\frac{1}{(1+Ts)^n}$	$\frac{t^{n-1} e^{-t/T}}{T^n (n-1)!}$

Laplace transform, $F(s)$	Time function, $f(t), t \geq 0$
$\frac{1}{s(1 + Ts)}$	$1 - e^{-t/T}$
$\frac{1}{s(1 + Ts)^2}$	$1 - \frac{t + T}{T} e^{-t/T}$
$\frac{\omega}{(s + a)^2 + \omega^2}$	$e^{-at} \sin \omega t$
$\frac{(s + a)}{(s + a)^2 + \omega^2}$	$e^{-at} \cos \omega t$
$\frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$	$1 + \frac{e^{-\zeta\omega_n t}}{\sqrt{1 - \zeta^2}} \sin(\omega_n \sqrt{1 - \zeta^2} t - \alpha)$ where $\cos \alpha = -\zeta$
$\frac{\omega_n^2}{s^2(s^2 + 2\zeta\omega_n s + \omega_n^2)}$	$t - \frac{2\zeta}{\omega_n} + \frac{1}{\omega_n \sqrt{1 - \zeta^2}} e^{-\zeta\omega_n t} \sin(\omega_n \sqrt{1 - \zeta^2} t - \theta)$ where $\theta = 2 \tan^{-1} \frac{\sqrt{1 - \zeta^2}}{-\zeta}$
$\frac{s}{(1 + Ts)(s^2 + \omega_n^2)}$	$\frac{-1}{(1 + T^2\omega_n^2)} e^{-t/T} + \frac{1}{\sqrt{1 + T^2\omega_n^2}} \cos(\omega_n t - \theta)$ where $\theta = \tan^{-1} \omega_n T$
$\frac{s}{(s^2 + \omega_n^2)^2}$	$\frac{1}{2\omega_n} t \sin \omega_n t$
$\frac{1}{(s + b)[(s + a)^2 + \omega^2]}$	$\frac{e^{-bt}}{(b - a)^2 + \omega^2} + \frac{e^{-at} \sin(\omega t - \theta)}{\omega[(b - a)^2 + \omega^2]^{1/2}}$ where $\theta = \tan^{-1} \frac{\omega}{b - a}$
$\frac{2abs}{[s^2 + (a + b)^2][s^2 + (a - b)^2]}$	$\sin at \sin bt$
$\frac{1 + as + bs^2}{s^2(1 + T_1 s)(1 + T_2 s)}$	$t + (a - T_1 - T_2) + \frac{b - aT_1 + T_1^2}{T_1 - T_2} e^{-t/T_1}$ $- \frac{b - aT_2 + T_2^2}{T_1 - T_2} e^{-t/T_2}$

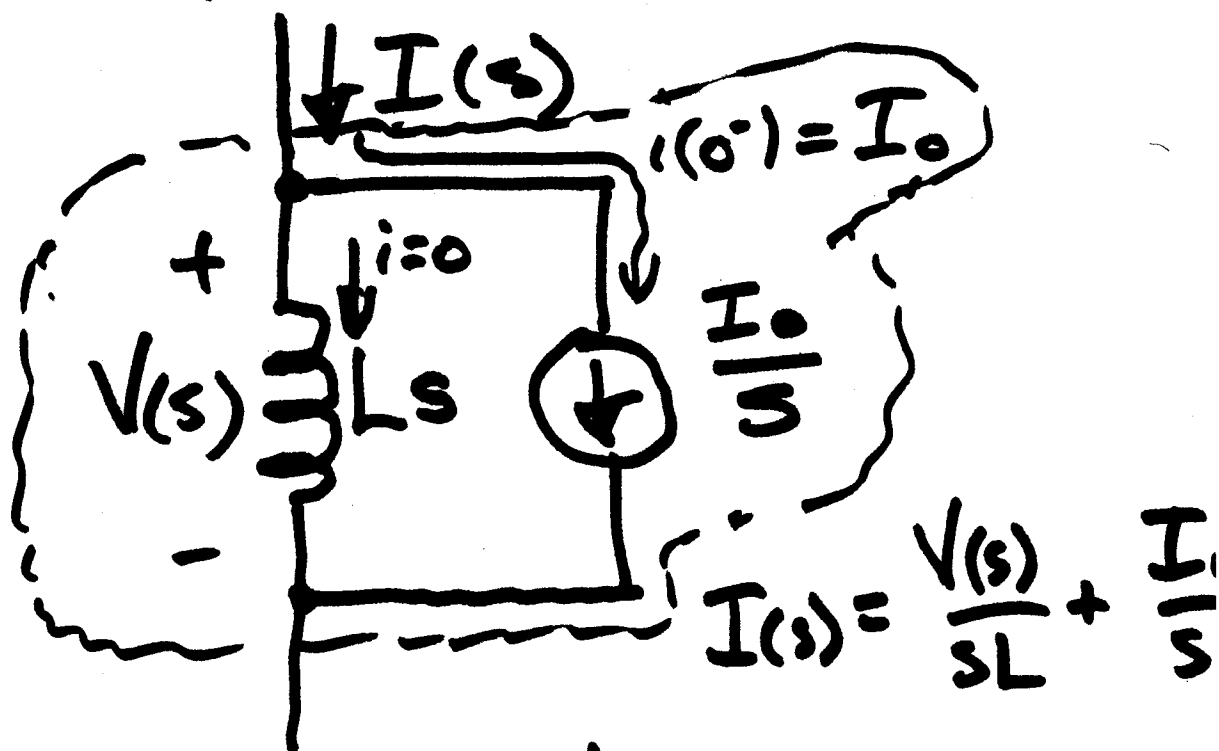
Laplace Review

$$s = \sigma + j\omega \quad (\text{complex freq})$$

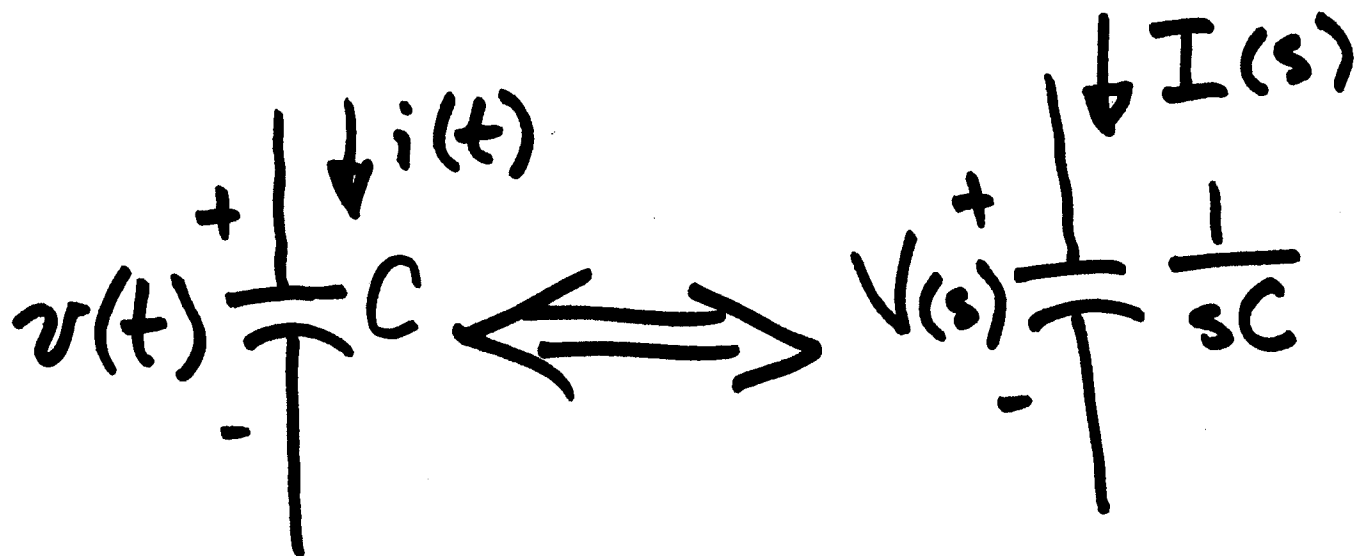


Initial Conditions:

if $i(0) = I_0$ then

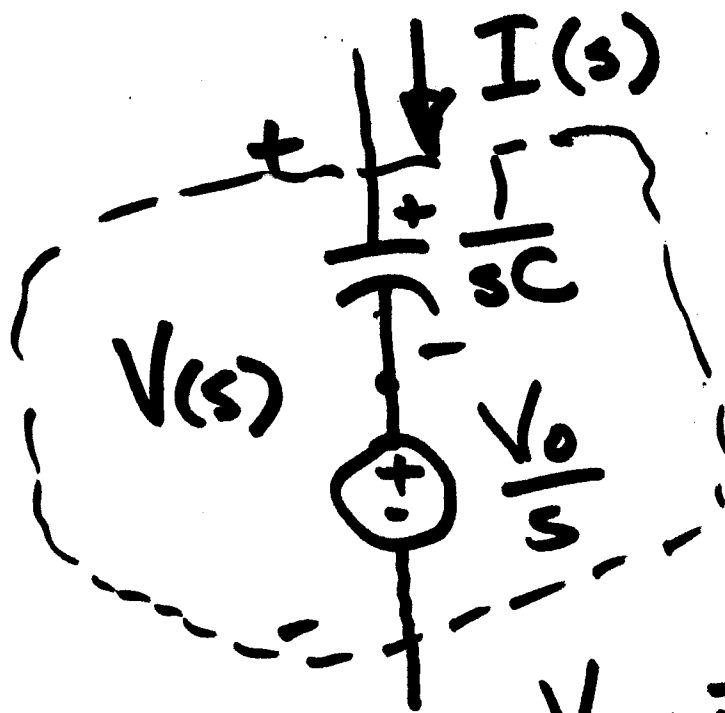


$$V(s) = sL \left(I(s) - \frac{I_0}{s} \right) = \underline{\underline{sL I(s) - L I_0}}$$



Initial Conditions:

if $v(0^-) = V_0$ then



$$V(s) = \frac{I(s)}{sC} + \frac{V_0}{s}$$

$$I(s) = sCV(s) - CV_0$$