

# LESSON 3: SAMPLING AND DISCRETE-TIME FOURIER ANALYSIS

- OBJECTIVES:
- DISCUSS DISCRETE-TIME SIGNALS & SYSTEMS
  - INTRODUCE SAMPLING THEORY
  - DISCRETE FOURIER TRANSFORM (DFT)

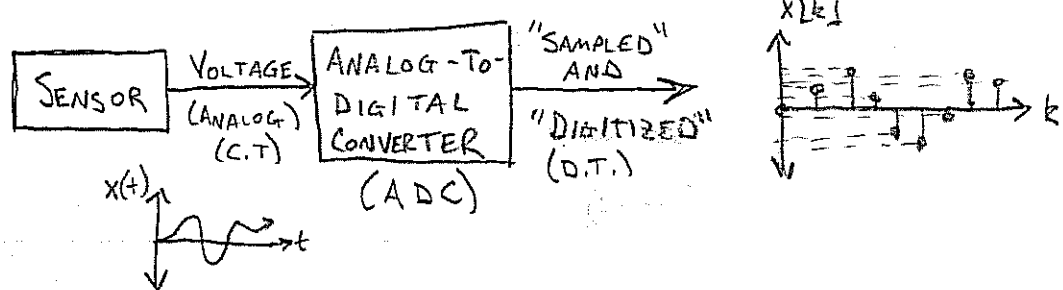
PREVIOUS LESSON DEALT WITH CONTINUOUS SIGNALS

- THESE REPRESENT REAL-WORLD PHENOMENA:

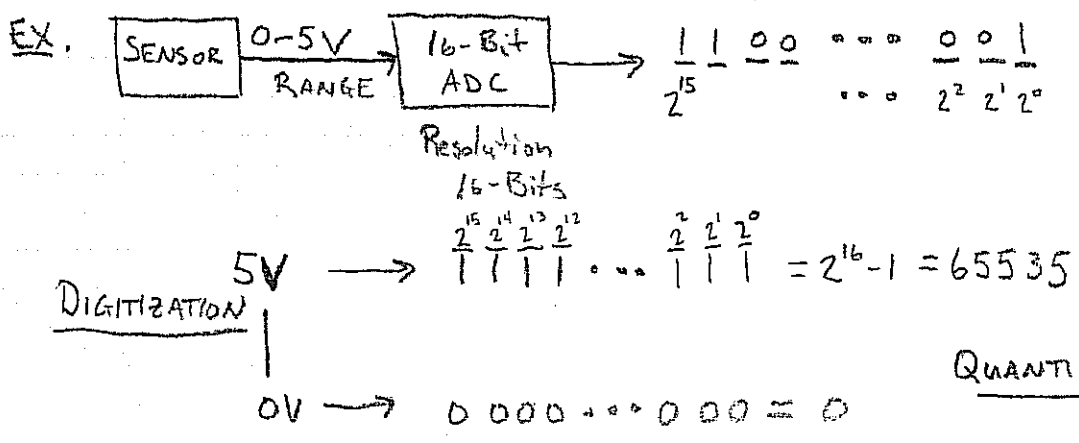
- ACCELERATION
- TEMPERATURE
- PRESSURE
- STRESS / STRAIN

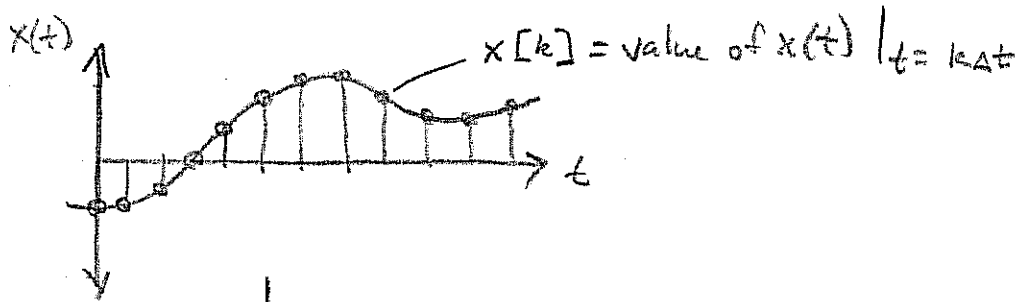
- DIGITAL COMPUTERS CANNOT HANDLE CONTINUOUS-DOMAIN SIGNALS!

## DISCRETE-TIME SIGNALS AND SYSTEMS



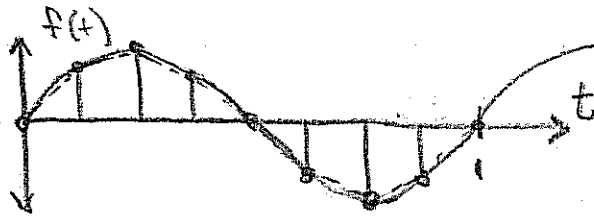
- SAMPLED IN TIME @ A FIXED INTERVAL,  $\Delta t$  (time-step, s)  
 $f_s = \frac{1}{\Delta t}$  (Hz)
- OUTPUT IS DIGITIZED FROM ANALOG VOLTAGE LEVEL TO DIGITAL NUMBER





↳ We can have a frequency-domain representation of  $x[k]$ : DFT

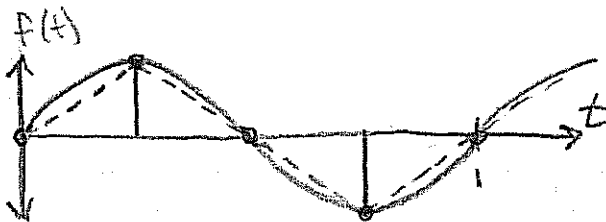
CONSIDER SIMPLE SINUSOID:



$$f(t) = \sin(2\pi t) \leftarrow f = 1 \text{ Hz}$$

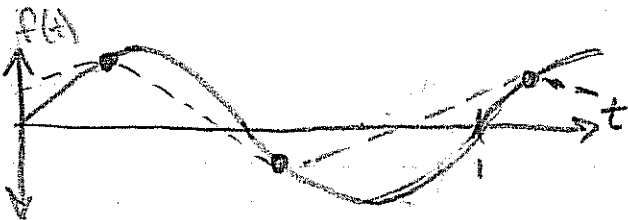
$$f_s = \text{SAMPLE RATE} = 8 \text{ Hz}$$

- Easy to infer true signal.



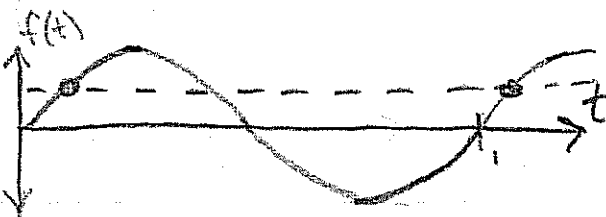
$$f_s = 4 \text{ Hz}$$

- Looks like triangular signal.



$$f_s = 2 \text{ Hz}$$

- Harder still to see true signal.



$$f_s = 1 \text{ Hz}$$

- LOOKS LIKE FLAT LINE (DC) SIGNAL!

\* ALIASING: HIGH-FREQ. SIGNALS APPEARING AS LOW-FREQ SIGNALS WHEN SAMPLED.

DISCRETE FOURIER TRANSFORM (DFT)

SAMPLED DATA CAN BE REPRESENTED IN THE FREQUENCY DOMAIN TOO.

CONSIDER A FINITE LENGTH DT SIGNAL:  $\{x_k\} = \{x_0, x_1, \dots, x_{N-1}\}^T$

$N = \#$  SAMPLES

$T = \text{Total Duration} = N(\Delta t)$

\* THE FINITE SIGNAL MIGHT BE THOUGHT OF AS PERIODIC,  
THEN WE CAN USE FOURIER SERIES:

FOURIER SERIES:  
(for CT signals)

$$X(t) = a_0 + 2 \sum_{k=1}^{\infty} \left( a_k \cos \frac{2\pi kt}{T} + b_k \sin \frac{2\pi kt}{T} \right)$$

$$a_k = \frac{1}{T} \int_0^T x(t) \cos \frac{2\pi kt}{T} dt$$

$$b_k = \frac{1}{T} \int_0^T x(t) \sin \frac{2\pi kt}{T} dt$$

→ USE COMPLEX NOTATION HERE AS WELL:

$$X_k = a_k - i b_k$$

EULER'S FORMULA:  $e^{-i(2\pi kt/T)} = \cos \frac{2\pi kt}{T} - i \sin \frac{2\pi kt}{T}$

$$X_k = \frac{1}{T} \int_0^T x(t) \left[ \cos \frac{2\pi kt}{T} - i \sin \frac{2\pi kt}{T} \right] dt$$

$$X_k = \frac{1}{T} \int_0^T x(t) e^{-i(2\pi kt/T)} dt$$

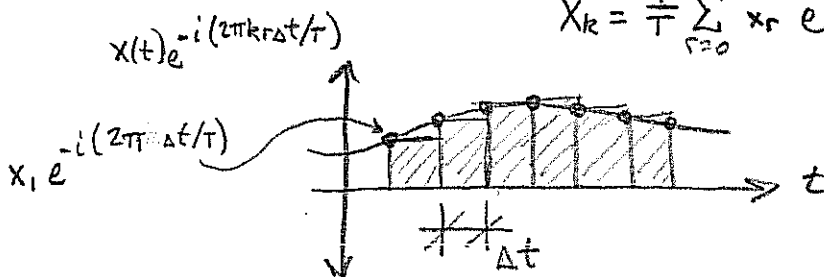
\* BUT THIS WAS FOR CT  $x(t)$ !

WHAT IF WE ONLY HAVE SAMPLED

INFORMATION ABOUT  $x(t) \Rightarrow x[k]$ ?

APPROXIMATE THE INTEGRAL:

$$X_k = \frac{1}{T} \sum_{r=0}^{N-1} x_r e^{-i(2\pi kr\Delta t/T)} \Delta t$$



$$T = N\Delta t \Rightarrow \frac{1}{N} = \frac{\Delta t}{T} \Rightarrow$$

$$X_k = \frac{1}{N} \sum_{r=0}^{N-1} x_r e^{-i(2\pi kr/N)}$$

$k = 0, 1, 2, \dots, N-1$

$x_r = r$ th sample of  $x[k]$

DISCRETE FOURIER TRANSFORM (DFT)

INVERSE  
DISCRETE  
FOURIER TRANSFORM  
(IDFT)

$$x_r = \sum_{k=0}^{N-1} X_k e^{i(2\pi k r/N)}$$

$r = 0, 1, 2, \dots, N-1$

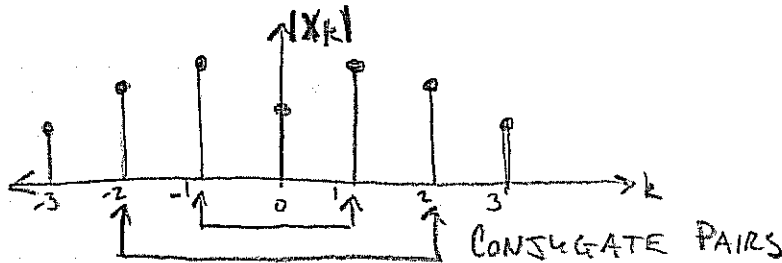
PROPERTIES OF DFT:

- INFORMATION LIMITED TO  $k=0$  TO  $k=N-1$
- FREQUENCIES:  $\omega_k = \frac{2\pi k}{T} = \frac{2\pi k}{N \Delta t}$  (rad/s)

• SYMMETRY (k=0):  $X_l = \frac{1}{N} \sum_{r=0}^{N-1} x_r e^{-i(2\pi r l/N)}$

$$X_{-l} = \frac{1}{N} \sum_{r=0}^{N-1} x_r e^{i(2\pi r l/N)}$$

$\rightarrow X_l = X_{-l}^*$  (COMPLEX-CONJUGATE)

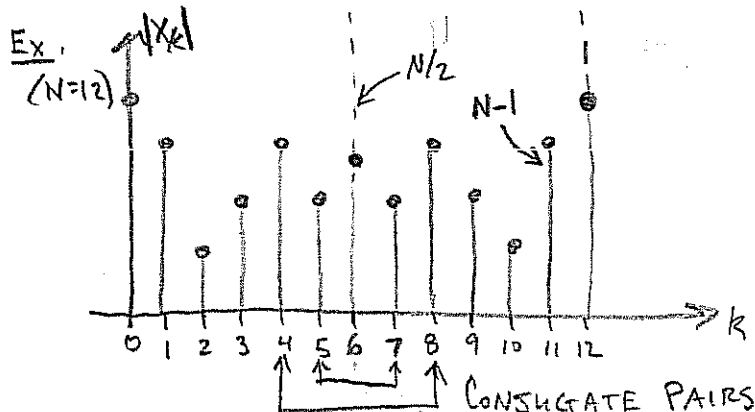


• SYMMETRY (k=N/2):  $X_1 = \frac{1}{N} \sum_{r=0}^{N-1} x_r e^{-i(2\pi 1 r/N)} = \frac{1}{N} \sum_{r=0}^{N-1} x_r e^{-i(\frac{2\pi r}{N})}$

$$X_{N-1} = \frac{1}{N} \sum_{r=0}^{N-1} x_r e^{-i(2\pi(N-1)r/N)} = \frac{1}{N} \sum_{r=0}^{N-1} x_r e^{+i(\frac{2\pi r}{N})} e^{-i(2\pi r)}$$

•  $X_{N-1} = X_1^*$  COMPLEX-CONJUGATE

ALSO:  $X_2 = X_{N-2}^*$   
 $X_3 = X_{N-3}^*$   
 $\vdots$   
 $X_{\frac{N}{2}-1} = X_{\frac{N}{2}+1}^*$



\* THE UNIQUE INFORMATION IS LIMITED TO

THE DOMAIN  $k=0$  TO  $k=\frac{N}{2}$

$$\rightarrow \omega_k \Big|_{k=\frac{N}{2}} = \frac{2\pi \cdot N}{2N \Delta t} = \frac{\pi}{\Delta t}$$

$$\omega_{\text{SAMPLE}} = \frac{2\pi}{\Delta t} \quad \therefore \quad \omega_{k=\frac{N}{2}} = \frac{1}{2} \omega_{\text{SAMPLE}}$$

$\therefore$  WE ONLY KNOW THE HARMONICS  
FROM DC ( $\omega=0$  rad/s) TO  $\frac{1}{2} \omega_{\text{SAMPLE}}$


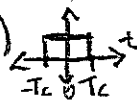
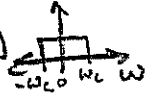

$\frac{1}{2} \omega_{\text{SAMPLE}} \text{ (rad./s)}$ $\frac{1}{2} f_{\text{SAMPLE}} \text{ (Hz)}$	}	NYQUIST FREQUENCY
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$\Rightarrow$  IN MATLAB USE FAST FOURIER TRANSFORM (FFT)

$\rightarrow$  Computationally efficient DTFT  
when  $N_{\text{SAMPLE}} = \text{power of } 2.$

$\rightarrow$  See homework.

\* FOURIER TRANSFORMS OF DISCRETE SIGNALS:

<u>FUNCTION</u>	<u>TIME-DOMAIN, <math>X[n]</math></u>	<u>FREQ. - DOMAIN, <math>X(\omega)</math></u>
IMPULSE 	$\delta[n] = \begin{cases} 1, & n=0 \\ 0, & n \neq 0 \end{cases}$	1
UNIT STEP	$u_s[n] = \begin{cases} 1, & n \geq 0 \\ 0, & n < 0 \end{cases}$	$\frac{1}{1 - e^{-j\omega}}$
RECT (time) 	$\text{rect}\left[\frac{(n - T_c/2)}{T_c}\right] = \begin{cases} 1, & -T_c/2 \leq n \leq T_c/2 \\ 0, & \text{otherwise} \end{cases}$	$\frac{\sin(\omega[T_c+1]/2)}{\sin(\omega/2)} e^{-j\omega T_c/2}$
COSINE	$\cos(an)$	$\pi[\delta(\omega-a) + \delta(\omega+a)]$
SINE	$\sin(an)$	$\pi/j[\delta(\omega-a) - \delta(\omega+a)]$
SINC = $\left(\frac{\sin x}{x}\right)$	$\text{sinc}(a+n)$	$e^{ja\omega}$
RECT (freq.) 	$\frac{W_c}{\pi} \text{sinc}\left(\frac{W_c}{\pi} n\right)$	$\text{RECT}\left[\frac{\omega}{2W_c}\right] = \begin{cases} 1, & -W_c/2 \leq \omega \leq W_c/2 \\ 0, & \text{otherwise} \end{cases}$
TRI (freq.) 	$\frac{W_c}{\pi} \text{sinc}^2\left(\frac{W_c}{\pi} n\right)$	$\text{TRI}\left[\frac{\omega}{2W_c}\right] = \begin{cases} \omega, & -W_c/2 \leq \omega \leq W_c/2 \\ 0, & \text{otherwise} \end{cases}$

\* PROPERTIES OF FOURIER TRANSFORMS OF DISCRETE SIGNALS:

<u>PROPERTY</u>	<u>TIME-DOMAIN</u>	<u>FREQ. - DOMAIN</u>
BASIC FUNCTION	$X[n]$	$X(\omega)$
TIME-SHIFT	$X[n-k]$	$e^{j\omega k} X(\omega)$
TIME-REVERSAL	$X[-n]$	$X(-\omega)$
FREQ. - SHIFT	$e^{j\omega_0 n} X[n]$	$X(\omega - \omega_0)$
DIFFERENTIATION (TIME-DOMAIN)	$\frac{d}{dt} X[n]$	$j\omega X(\omega)$
MODULATION	$X[n] \cos \omega_0 n$	$\frac{1}{2} [X(\omega + \omega_0) + X(\omega - \omega_0)]$
CONVOLUTION (TIME-DOMAIN)	$X_1[n] \otimes X_2[n]$ $\left( \sum_{k=-\infty}^{\infty} X_1[k] X_2[n-k] \right)$	$X_1(\omega) X_2(\omega)$

\* NOTE: DFT IS FREQUENCY SAMPLED VERSION OF F.T.,  $X(\omega)$