

LESSON 4: ALIASING AND FILTERING

- OBJECTIVES: - UNDERSTAND How ALIASING DISTORTS SIGNALS
 - LEARN ABOUT FILTERING
 - UNDERSTAND LEAKAGE

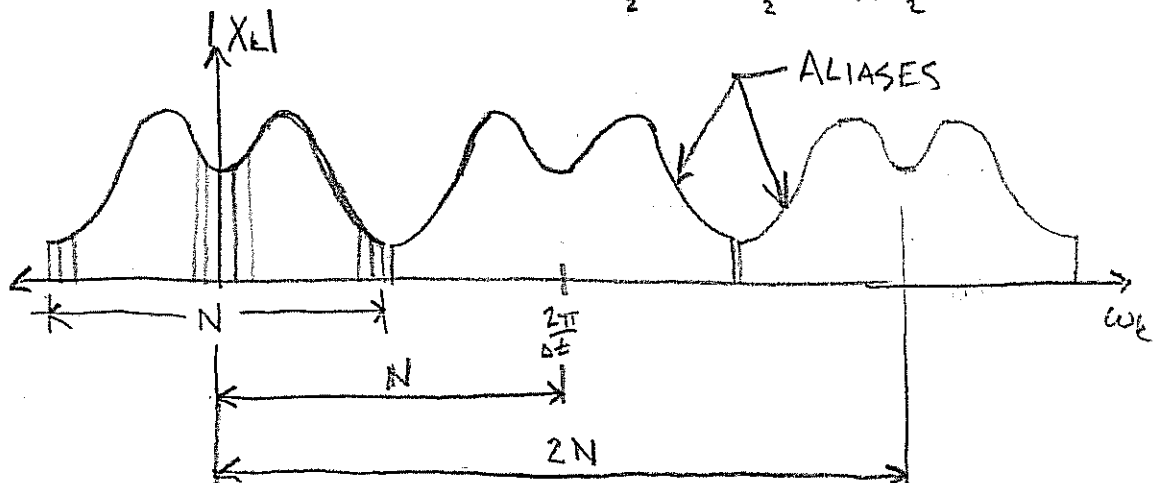
$$\text{IDFT: } X_k = \frac{1}{N} \sum_{r=0}^{N-1} x_r e^{-i(2\pi kr/N)}$$

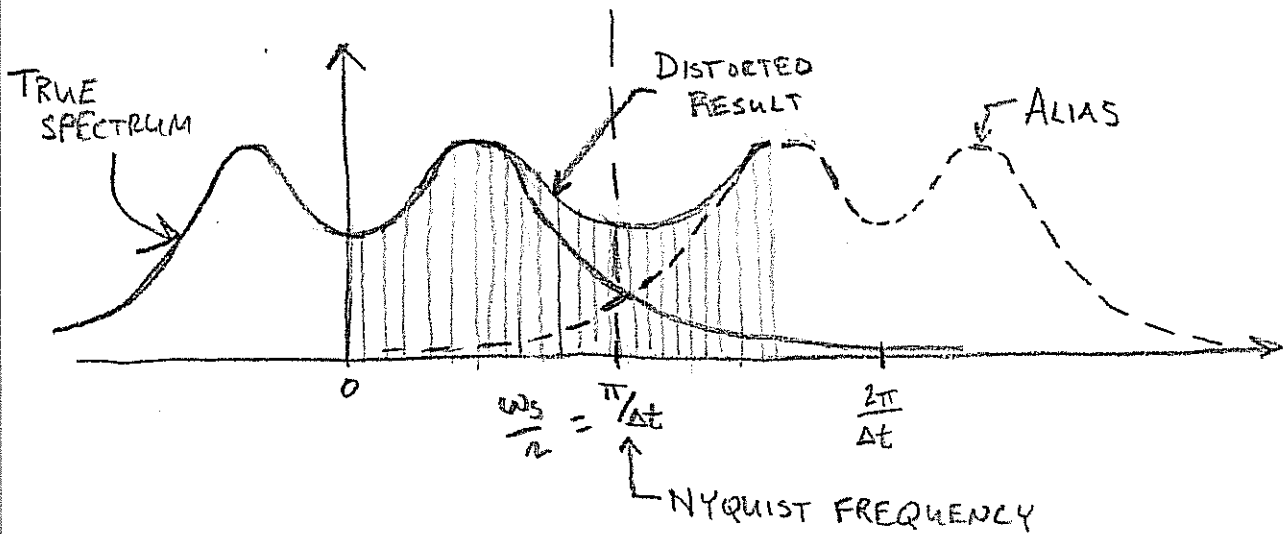
- LAST LESSON: We saw that the unique information in the DFT was limited to a frequency range of $0 \rightarrow \frac{\omega_{\text{SAMPLE}}}{2}$.

- Outside this range we only get complex conjugate symmetry
 - What happens to higher frequency info?

$$\begin{aligned} \text{CONSIDER: } X_{N+l} &= \frac{1}{N} \sum_{r=0}^{N-1} x_r e^{-i(2\pi r(N+l)/N)} \\ &= \frac{1}{N} \sum_{r=0}^{N-1} x_r e^{-i(2\pi rl/N)} e^{-i2\pi r} \\ &= X_l \end{aligned}$$

$$\begin{aligned} \therefore X_{N+l} &= X_l \rightarrow X_N = X_0 \\ X_{2N} &= X_N = X_0 \\ X_{\frac{N}{2}} &= X_{N+\frac{N}{2}} = X_{2N+\frac{N}{2}} \end{aligned}$$





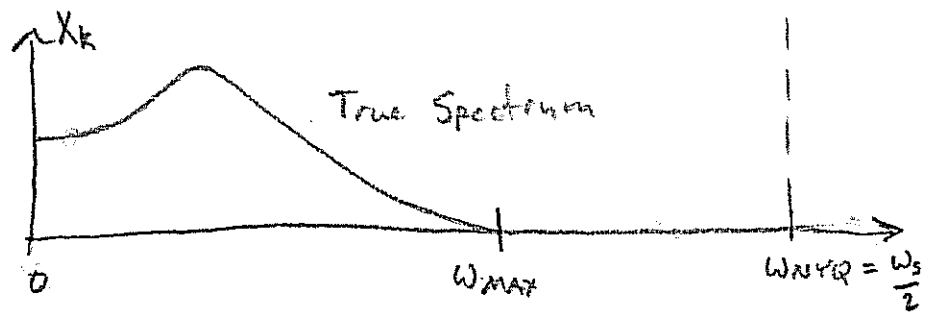
* WHEN SIGNIFICANT SPECTRAL COMPONENTS EXIST PAST NYQUIST FREQUENCY, THEY DISTORT THE DFT.

- THESE EFFECTS CANNOT BE REMOVED!

OVERCOMING ALIASING

1) SAMPLING RATE SELECTION:

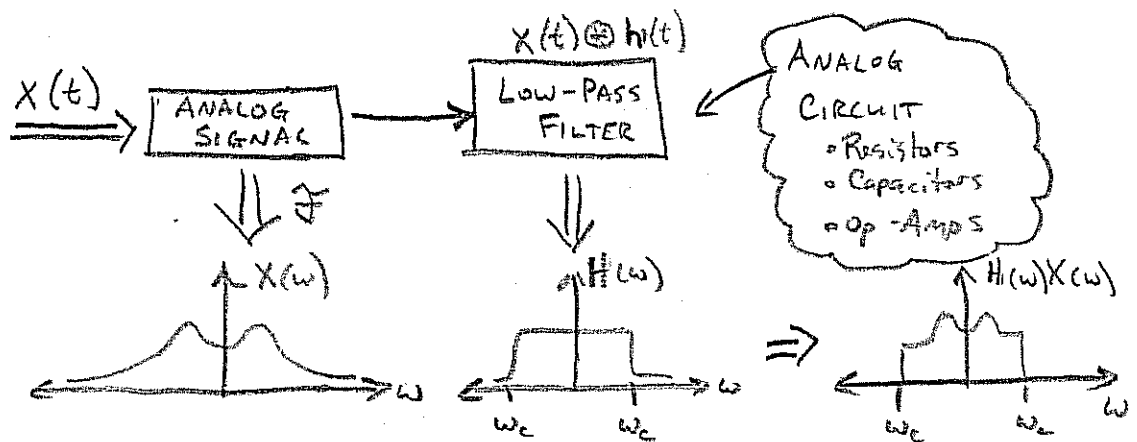
IF POSSIBLE, CHOOSE $\omega_{\text{SAMPLE}} \approx 3 \times$ HIGHEST FREQUENCY HARMONIC



$$\omega_{\text{NYQ}} = \frac{3}{2} \omega_{\text{MAX}}$$

2) ANTI-ALIASING FILTERS

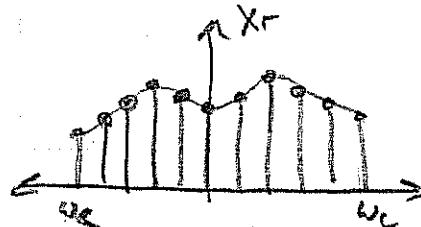
REMOVE HIGH-FREQUENCY CONTENT OF SIGNAL
BEFORE IT IS SAMPLED.



ω_c = Cut-off frequency

select to be $\leq \frac{\omega_{\text{sample}}}{2}$

SAMPLE FILTERED DATA:



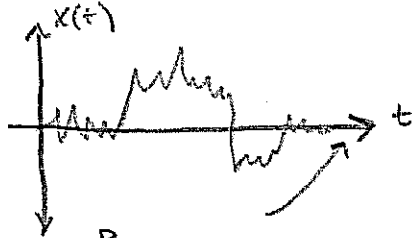
IF $\omega_c < \frac{\omega_{\text{SAMPLE}}}{2}$

NO ALIASING

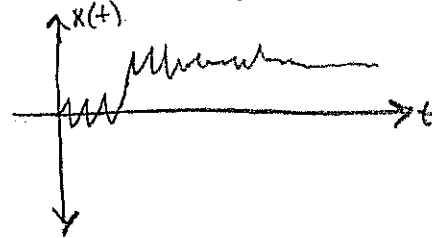
* ADDITIONAL ISSUES MAY LEAD TO
FREQUENCY-DOMAIN DISTORTIONS...

PROBLEMS WITH OFFSETS AND DISCONTINUITIES:

- DFT DERIVED ASSUMING PERIODICITY

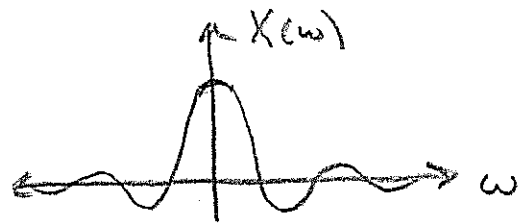
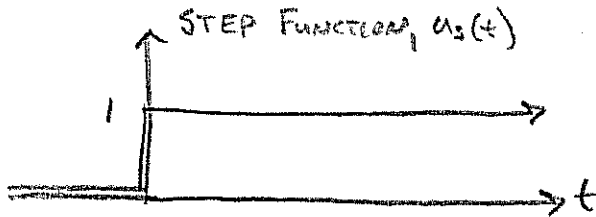


RETURNS TO ZERO → OK

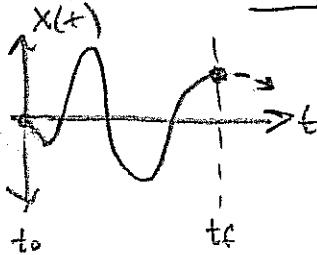


PERMANENT OFFSET: LEADS TO "LEAKAGE"

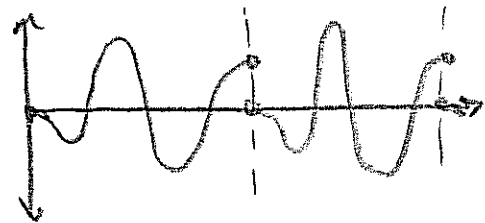
* OFFSET LIKE A STEP:



* TIME-DOMAIN WINDOWING CAUSES SIMILAR DISTORTIONS

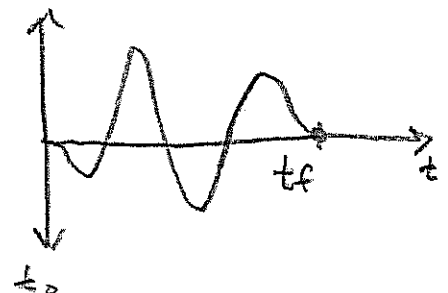
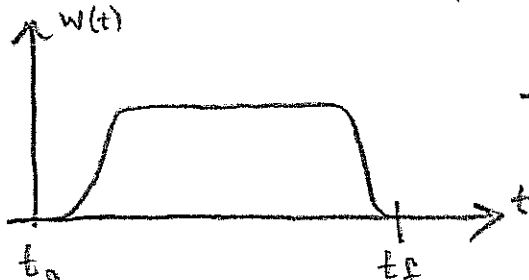


DFT Assumes



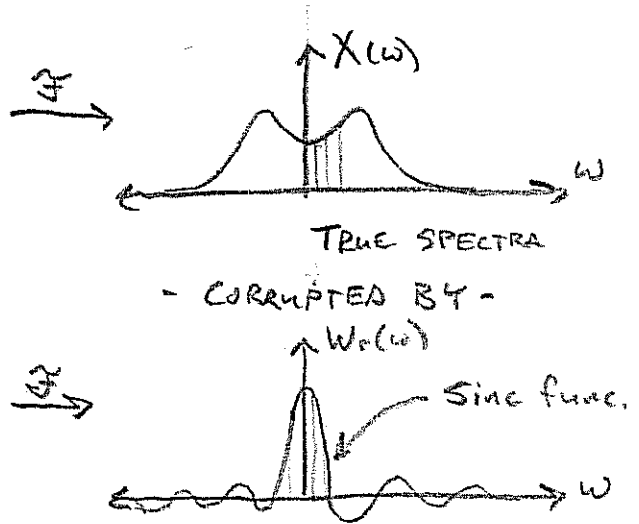
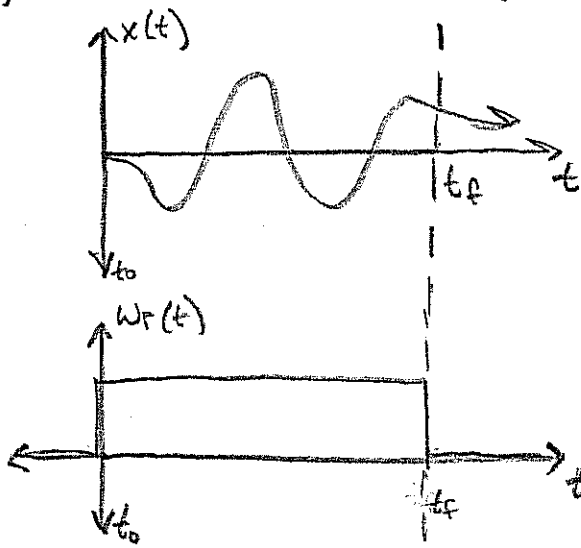
DISCONTINUITY CAUSES STEP-LIKE RESPONSE IN FREQ. DOMAIN

We might change our window shape to fix this leakage issue:

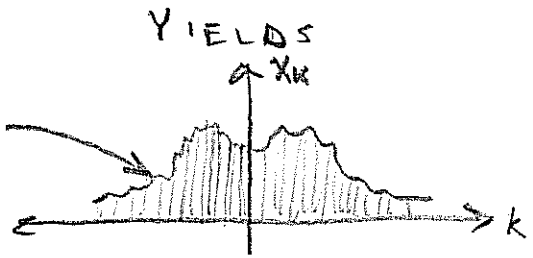


WINDOW OPTIONS :

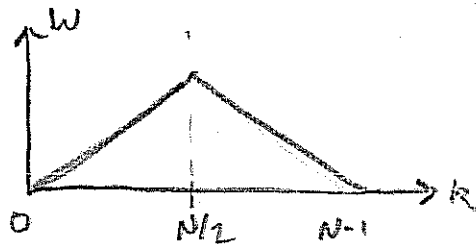
1) RECTANGULAR WINDOW:



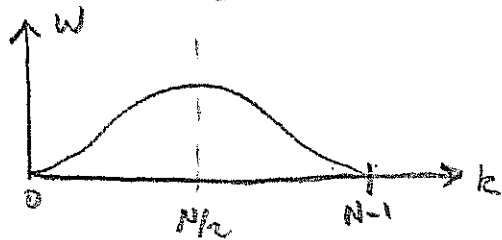
- DISTORTED VERSION
- LESS PROBLEMATIC AS T IS INCREASED
- LEARN THE MECHANISM NEXT LECTURE



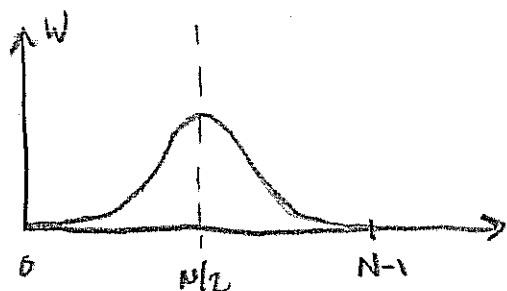
2) TRIANGULAR WINDOW (BARTLETT)



3) HANNING:

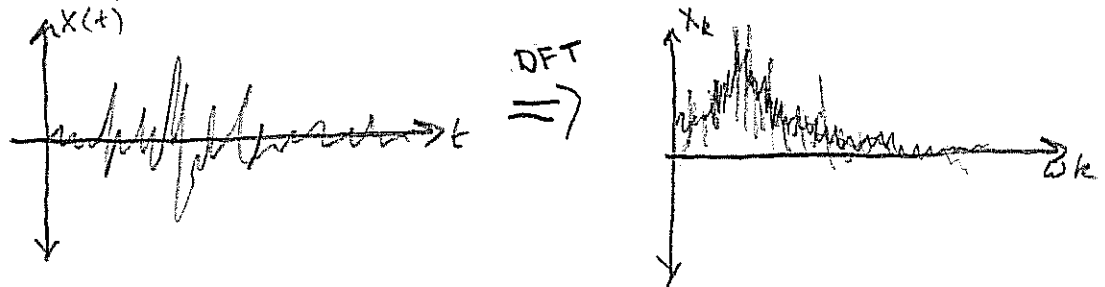


4) HAMMING:



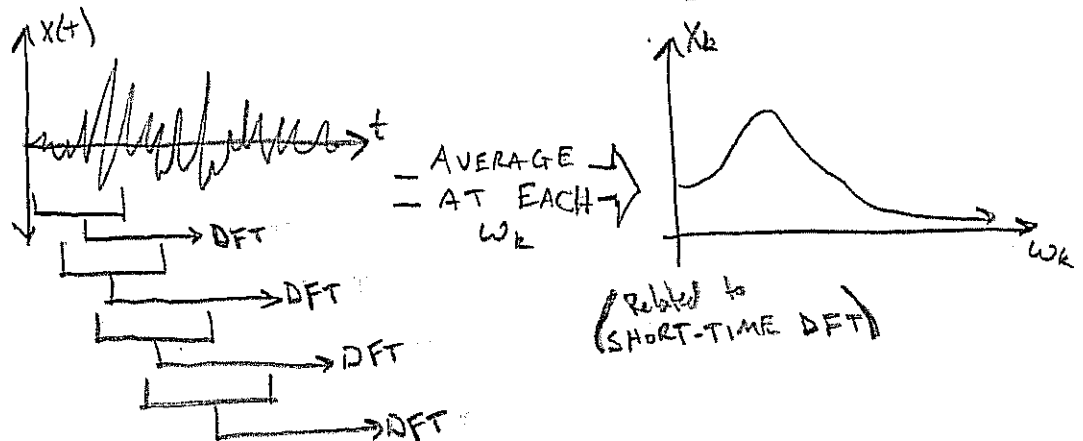
PROBLEMS WITH NOISE:

- REAL DATA IS VERY NOISY:



- IT WOULD BE NICE TO SMOOTH IT:

◦ ONE APPROACH TO ACCOMPLISH THIS IS TO PERFORM MULTIPLE DFTS OVER SHORT SUBSETS OF DATA WITH DIFFERING OFFSETS



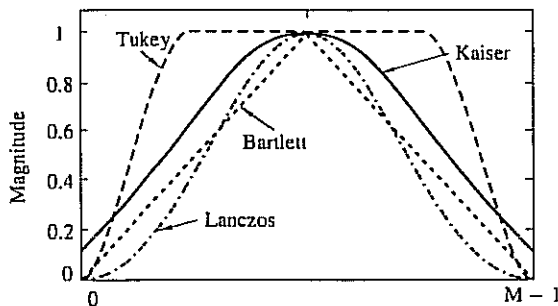
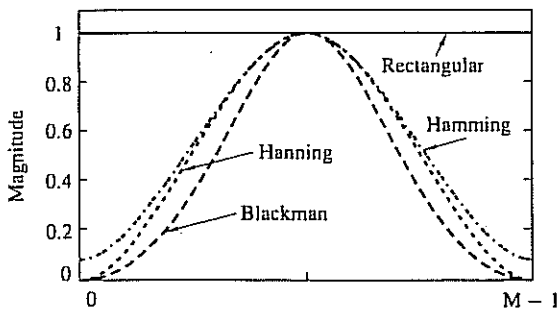
◦ LIMITS TO THE EFFECTIVENESS OF THIS APPROACH

◦ NEEDS LOTS OF DATA

◦ PRONE TO LEAKAGE (OVER AVERAGING)

TABLE 8.1 WINDOW FUNCTIONS FOR FIR FILTER DESIGN

Name of window	Time-domain sequence, $h(n), 0 \leq n \leq M-1$
Bartlett (triangular)	$1 - \frac{2 \left n - \frac{M-1}{2} \right }{M-1}$
Blackman	$0.42 - 0.5 \cos \frac{2\pi n}{M-1} + 0.08 \cos \frac{4\pi n}{M-1}$
Hamming	$0.54 - 0.46 \cos \frac{2\pi n}{M-1}$
Hanning	$\frac{1}{2} \left(1 - \cos \frac{2\pi n}{M-1} \right)$
Kaiser	$\frac{I_0 \left[\alpha \sqrt{\left(\frac{M-1}{2} \right)^2 - \left(n - \frac{M-1}{2} \right)^2} \right]}{I_0 \left[\alpha \left(\frac{M-1}{2} \right) \right]}$
Lanczos	$\left\{ \frac{\sin \left[2\pi \left(n - \frac{M-1}{2} \right) / (M-1) \right]}{2\pi \left(n - \frac{M-1}{2} \right) / \left(\frac{M-1}{2} \right)} \right\}^L$ $L > 0$
Tukey	$\frac{1}{2} \left[1 + \cos \left(\frac{n - (1+\alpha)(M-1)/2}{(1-\alpha)(M-1)/2} \pi \right) \right]$ $\alpha(M-1)/2 \leq \left n - \frac{M-1}{2} \right \leq \frac{M-1}{2}$



SOURCE :

Proakis & Manolakis

"DIGITAL SIGNAL PROCESSING:

Principles, Algorithms, and Applications," 3rd Ed.

Prentice Hall (1996)